

# Dynamic System Identification for Robot-Assisted Transcranial Magnetic Stimulation

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**ABSTRACT:** Robotic-assisted Transcranial magnetic stimulation has been technically developed and widely used to enhance the long-term outcome of brain stimulation. In this paper, the overall dynamic model of the robotic Transcranial stimulation has been successfully analysed; nevertheless, its parameter estimation leads to a significant challenge due to the complicated dynamic nature of the system. To solve this limitation, system identification based on auto-regressive moving average with an exogenous signal model was consequently implemented. The predicted model was carried out and shown to be an effective data matching with the best-fit percentage of 97.10 %. The optimized gains of proportional and integral control were simulatively achieved using Ziegler–Nichols tuning method incorporated with the transfer function of the proposed dynamic model. These algorithm variables were subsequently used in the robotic implicit force control to regulate the contact forces, in which the human’s head is tracked by a 3D scanner in real-time. The results demonstrate significant satisfaction with the overall qualitative performance of the robotic control, and it simplifies effective tracking of the contact force target by indicating closely the desired input with root mean square error of 0.014 N. Hence, it can be concluded that the implicit force control associated with the human’s head detection system, is considered acceptable for the effective Transcranial magnetic stimulation to facilitate safe and reliable physical collaboration between the human and robot.

**KEYWORDS:** Implicit Force Control; Dynamic Model Identification; Robotic Transcranial magnetic stimulation

## INTRODUCTION

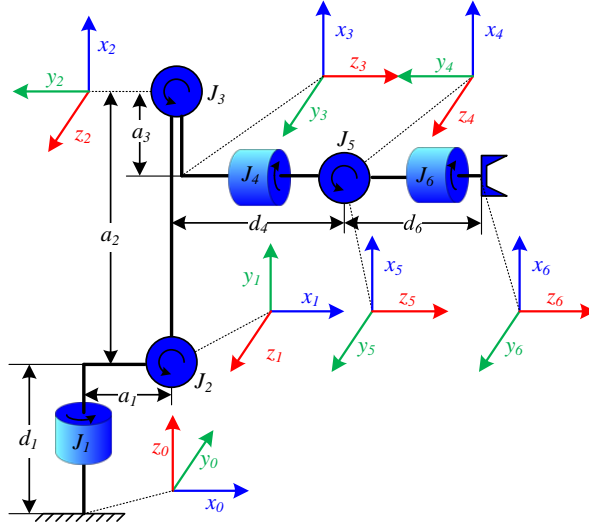
Transcranial magnetic stimulation (TMS) is a non-invasive, painless and powerful technique of brain stimulation using magnetic field generated by the electric current at a specific area of the brain. TMS, which has been developed and used for more than 30 years, is widely used to study human brain physiology, particularly in human motor functions and the pathophysiology of brain disorders. A TMS coil is an electrical pulse generator, which is placed to a patient’s scalp to stimulate the neural elements. The repetitive TMS effects to motor function recovery of Parkinson’s disease, stroke and movement disorder [1]. Traditionally, the TMS coil is held and positioned manually by a specialist doctor without precise clinical experiments. This leads to promising applications of robotics TMS systems to enhance more effective accuracy and reliable TMS.

Robot-assisted TMS systems have been developed by a number of robotic researchers. However, in order to adopt optimized force control in the non-linear robotic TMS system, it required the TMS plant transfer function [2]. The mathematical model computation is more complicated when relevant model parameters are estimated i.e. some impedance parameters of robots, force sensors, robot grippers and human environment [3-4]. To overcome the problem, system identification, which plays a crucial role to deal with the determination of overall system dynamic characteristics using applicable test results, can be strategically implemented. Therefore, this paper describes how to construct an estimated plant transfer function of a robot-assisted TMS system using MATLAB System Identification Toolbox. This technique involves in the optimal test design to effectively deliver the relative information data for an appropriate TMS model fitting. The transfer function of the TMS robot is subsequently evaluated before being utilized to further develop an optimized implicit force controller based on Proportional-Integral (PI) control. Moreover, the comparison between the simulation and real experimental results are finally conveyed.

## SYSTEM ANALYSIS AND MODELLING OF ROBOTIC TMS SYSTEM

Conceptual Design and Test Procedure of Robotic TMS System




**Figure 2.** Manipulator kinematic chain and D-H parameters

**Table 1.** DH parameter estimations for the KUKA robot

Joint	$\theta^\circ$	$d(\text{mm})$	$a(\text{mm})$	$\alpha^\circ$	$\theta_{\min}^\circ$	$\theta_{\max}^\circ$
$J_1$	$\theta_1$	$d_1=675$	$a_1=260$	-90	-185	185
$J_2$	$\theta_2$	$d_2=0$	$a_2=680$	0	-155	35
$J_3$	$\theta_3-90^\circ$	$d_3=0$	$-a_3=-35$	-90	-130	154
$J_4$	$\theta_4$	$d_4=670$	$a_4=0$	90	-350	350
$J_5$	$\theta_5$	$d_5=0$	$a_5=0$	-90	-130	130
$J_6$	$\theta_6$	$d_6=115$	$a_6=0$	0	-350	350

The KUKA robot joints are named as  $J_1$  to  $J_6$ , and the lengths of the robot linkages are denoted as  $a_i$  and  $d_i$ . The transformation matrices ( $T_i^{i+1}$ ) of the robot can be calculated as explained in the following equations, in which the following notation is used:  $Rot_{z,\theta_i}$  and  $Rot_{x,\alpha_i}$  denote the rotation with  $\theta_i$  and  $\alpha_i$  around  $z$  and  $x$  axes. The terms  $Trans_{z,d_i}$  and  $Trans_{x,a_i}$  stand for translation with  $d_i$  and  $a_i$  along  $x$  and  $z$  axes respectively. The parameters:  $c_i$  and  $s_i$  represent  $\cos(\theta_i)$  and  $\sin(\theta_i)$ , and then the transformation matrices can be presented as:

$$T_i^{i+1} = Rot_{z,\theta_i} \cdot Trans_{z,d_i} \cdot Trans_{x,a_i} \cdot Rot_{x,\alpha_i} \quad (1)$$

$$T_0^1 = Rot_{z,\theta_1} \cdot Trans_{a_1,0,d_1} \cdot Rot_{x,-90^\circ}, T_1^2 = Rot_{z,\theta_2} \cdot Trans_{a_2,0,0}, T_2^3 = Rot_{z,\theta_3-90^\circ} \cdot Trans_{-a_3,0,0} \cdot Rot_{x,-90^\circ} \quad (2)$$

$$T_3^4 = Rot_{z,\theta_4} \cdot Trans_{0,0,d_4} \cdot Rot_{x,90^\circ}, T_4^5 = Rot_{z,\theta_5} \cdot Rot_{x,-90^\circ} \text{ and } T_5^6 = Rot_{z,\theta_6} \cdot Trans_{0,0,d_6}$$

$$\begin{aligned}
 T_0^1 &= \begin{bmatrix} C_1 & 0 & -S_1 & a_1C_1 \\ S_1 & 0 & C_1 & a_1S_1 \\ 0 & -1 & 0 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}, & T_1^2 &= \begin{bmatrix} C_2 & -S_2 & 0 & a_2C_2 \\ S_2 & C_2 & 0 & a_2S_2 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}, & T_2^3 &= \begin{bmatrix} S_3 & 0 & C_3 & -a_3S_3 \\ -C_3 & 0 & S_3 & a_3C_3 \\ 0 & -1 & 0 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}, \\
 T_3^4 &= \begin{bmatrix} C_4 & 0 & S_4 & 0 \\ S_4 & 0 & -C_4 & 0 \\ 0 & 1 & 0 & d_4 \\ 0 & 0 & 0 & 1 \end{bmatrix}, & T_4^5 &= \begin{bmatrix} C_5 & 0 & -S_5 & 0 \\ S_5 & 0 & C_5 & 0 \\ 0 & -1 & 0 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}, & T_5^6 &= \begin{bmatrix} C_6 & -S_6 & 0 & 0 \\ S_6 & C_6 & 0 & 0 \\ 0 & 0 & 1 & d_6 \\ 0 & 0 & 0 & 1 \end{bmatrix}.
 \end{aligned} \tag{3}$$

The homogenous transformation from the robot base coordination to the robot end effector is formulated by:

$$T_1^6 = T_0^1 T_1^2 T_2^3 T_3^4 T_4^5 T_5^6 = \begin{bmatrix} r_{11} & r_{12} & r_{13} & p_x \\ r_{21} & r_{22} & r_{23} & p_y \\ r_{31} & r_{32} & r_{33} & p_z \\ 0 & 0 & 0 & 1 \end{bmatrix} \tag{4}$$

where,  $r_{i1}$ ,  $r_{i2}$  and  $r_{i3}$  are orthonormal orientation vectors of x-y-z axes respectively, and p is a position vector. The inverse kinematics of the robot used to transform motion specifications into corresponding joint space can be defined as the determination of each robot joint regarding the position and orientation of the robot end effector. Two techniques of the inverse kinematic solution are made up of mathematical or algebraic (always implemented with higher DOF robots) and geometrical approaches. Thus, the geometrical method was appropriately used to solve the KUKA KR-16 inverse kinematic problem and it gives:

$$\theta_1 = \text{Atan2}(p_x, p_y) - \text{Atan2}\left(d_3, \pm \sqrt{p_x^2 + p_y^2 - d_3^2}\right) \tag{5}$$

$$\theta_2 = \text{Atan2}\left\{ \begin{aligned} & [(-a_3 - a_3c_3)p_z - (c_1p_x + s_1p_y)(d_4 - a_2s_3)], \\ & [(a_2s_3 - d_4)p_z - (a_3 + a_2c_3) \cdot (c_1p_x + s_1p_y)] \end{aligned} \right\} - \theta_3 \tag{6}$$

$$\theta_3 = \text{Atan2}(a_3, d_4) - \text{Atan2}\left(a_3c_3 - d_4s_3, \pm \sqrt{a_3^2 + d_4^2 + (a_3c_3 - d_4s_3)^2}\right) \tag{7}$$

$$\theta_4 = \text{Atan2}[-(r_{13}s_1 + r_{23}c_1), (-r_{13}c_1c_{23} - r_{23}s_1c_{23} + r_{33}s_{23})]$$

$$s_{23} = \frac{(-a_3 - a_2c_3)p_z + (c_1p_x + s_1p_y)(a_2s_3 - d_4)}{p_z^2 + (c_1p_x + s_1p_y)^2} \tag{8}$$

$$c_{23} = \frac{(a_2s_3 - d_4)p_z - (a_3 + a_2c_3)(c_1p_x + s_1p_y)}{p_z^2 + (c_1p_x + s_1p_y)^2}$$

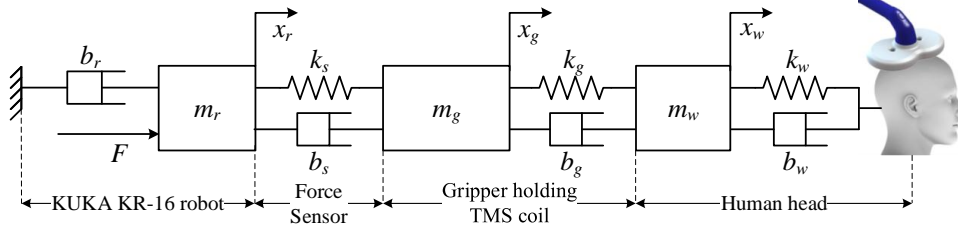
$$\theta_5 = \text{Atan2}(s_5, c_5) \tag{9}$$

$$\theta_6 = \text{Atan2}(s_6, c_6) \tag{10}$$

To analyse the relation between the robot Cartesian velocity and joint space velocity, the Jacobian matrix or  $J(q)$ , representing the matrix of all first-order partial derivatives of a vector-valued function, was implemented. Therefore, the robot manipulator Jacobian  $\xi(t)$  can be expressed as in Equation (11). When the Jacobian is given as a square matrix, then  $\dot{q}$  can be determined as in Equation (12)

$$\xi = \begin{bmatrix} v_n^0 \\ \omega_{0,n}^0 \end{bmatrix} = J(q) \frac{dq}{dt} = [J_1(q), J_2(q), \dots, J_n(q)] \begin{bmatrix} \dot{q}_1 \\ \vdots \\ \dot{q}_n \end{bmatrix} \quad (11)$$

$$\dot{q} = J^{-1}(q)\xi \quad (12)$$



**Figure 3.** Second-order-lumped-mass model of the Robotic-assisted TMS system

The following notation is used:  $q_1$  to  $q_n$  are the robot joint variables,  $\omega_{0,n}^0$  is the angular velocity of the end effector and  $v_n^0$  is a linear velocity of the end effector. This finding is subsequently used to compute the velocity of the robot end effector of any point in the robot real-time path control, which is addressed in the robotic implicit force control implementation for TMS section.

#### Dynamic Mechanical Model of Robotic TMS

This section explains a mathematical dynamic model of the robotic TMS system. Initially, the system is assumed to be sub-equivalent-lumped-mass systems, illustrated in Figure 3. The KUKA KR-16 robot can be modelled by the rigid body model ( $m_r$ ) associated with effective viscous damping ( $b_r$ ) to ground. The robot itself can be driven by force  $F$  generated by robot force control. The multi-axis force sensor is represented as a spring and dashpot system, in which the stiffness and viscous damping are  $k_s$  and  $b_s$  respectively. The total mass, stiffness and damping factor of the robot gripper while holding the rigid-body TMS coil are  $m_g$ ,  $k_g$  and  $b_g$ . The human model is confined to the environment component, which is contacted by the stimulating coil. The human mass represents as  $m_w$ , and the viscous damping and stiffness are  $b_w$  and  $k_w$ . In addition, the displacements of the masses ( $m_r$ ,  $m_g$  and  $m_w$ ) are respectively defined as  $x_r$ ,  $x_g$  and  $x_w$ .

There are many ways to establish state variables to describe a system space. By applying Newton's 2<sup>nd</sup> law of motion, the system dynamic equations of the rigid body masses of  $m_r$ ,  $m_g$  and  $m_w$  are computed. The design of the state space for the dynamic model of the TMS specifically requires the characteristic Equations (17-22). Then, the plant transfer functions of the system can be examined using modelling presented in the state space. This section describes how the transfer functions of a series of lumped-parameter models of the robotic TMS system are examined. The state-space equations based modelling were formulated and given in the following equations.

$$\frac{dx_r(n)}{dt} = \dot{x}_r(n) \quad (13)$$

$$\ddot{x}_r(n) = \frac{F(n)}{m_r} - \frac{(b_r + b_s)\dot{x}_r(n)}{m_r} + \frac{b_s\dot{x}_g(n)}{m_r} - \frac{k_s x_r(n)}{m_r} + \frac{k_s x_g(n)}{m_r} \quad (14)$$

$$\frac{dx_g(n)}{dt} = \dot{x}_g(n) \quad (15)$$

$$\ddot{x}_g(t) = \frac{b_s \dot{x}_r}{m_g} - \frac{(b_s + b_g)\dot{x}_g}{m_g} + \frac{b_g \dot{x}_w}{m_g} + \frac{k_s x_r}{m_g} - \frac{(k_s + k_g)x_g}{m_g} + \frac{k_g x_w}{m_g} \quad (16)$$

$$\frac{dx_w(n)}{dt} = \dot{x}_w(n) \quad (17)$$

$$\ddot{x}_w(t) = \frac{b_g \dot{x}_g}{m_w} - \frac{(b_g + b_w) \dot{x}_w}{m_w} + \frac{k_g x_g}{m_w} - \frac{(k_g + k_w) x_w}{m_w} \quad (18)$$

$$\dot{X}(t) = AX(t) + BF(t) \quad (19)$$

$$Y(t) = CX(t) + DF(t) \quad (20)$$

where  $X$  is called the state vector,  $Y$  is called the output vector,  $F$  is called the input vector,  $A$  is called the state matrix,  $B$  is called the input matrix,  $C$  is called the output matrix, and finally,  $D$  is called the feed-through matrix [9-10]. According to the TMS system, vector-matrix and system output forms can be written in the state space representation as:

$$\frac{d}{dt} \begin{bmatrix} x_r \\ \dot{x}_r \\ x_g \\ \dot{x}_g \\ x_w \\ \dot{x}_w \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & 0 & 0 & 0 \\ \frac{k_s}{m_r} & -\frac{(b_r + b_s)}{m_r} & \frac{k_s}{m_r} & \frac{b_s}{m_r} & 0 & 0 \\ 0 & 0 & 0 & 1 & 0 & 0 \\ \frac{k_s}{m_g} & \frac{b_s}{m_g} & -\frac{(k_g + k_s)}{m_g} & -\frac{(b_g + b_s)}{m_g} & \frac{k_g}{m_g} & \frac{b_g}{m_g} \\ \frac{1}{m_g} & \frac{1}{m_g} & \frac{1}{m_g} & \frac{1}{m_g} & \frac{1}{m_g} & \frac{1}{m_g} \\ 0 & 0 & 0 & 0 & 0 & 1 \\ 0 & 0 & \frac{k_g}{m_w} & \frac{b_g}{m_w} & -\frac{(k_g + k_w)}{m_w} & -\frac{(b_g + b_w)}{m_w} \end{bmatrix} \begin{bmatrix} x_r \\ \dot{x}_r \\ x_g \\ \dot{x}_g \\ x_w \\ \dot{x}_w \end{bmatrix} \quad (21)$$

$$+ \begin{bmatrix} 0 & 0 & 0 & 0 & 0 & 0 \\ \frac{1}{m_r} & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} F \\ 0 \\ 0 \\ 0 \\ 0 \\ 0 \end{bmatrix} \quad \text{and, } \begin{bmatrix} 1 & 0 & 0 & 0 & 0 & 0 \\ 0 & 0 & 1 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 & 1 & 0 \end{bmatrix} \begin{bmatrix} x_r \\ \dot{x}_r \\ x_g \\ \dot{x}_g \\ x_w \\ \dot{x}_w \end{bmatrix} + [0]F$$

The following equations explain how the system transfer functions are represented based on the proposed state-space form. Consider the state space system is shown in Equations (26) and (27). Ogata (2002) mentioned that the transfer matrix,  $G(s)$ , which relates to the output vector  $Y(s)$  and the input vector  $F(s)$  can be derived as [11]:

$$Y(s) = G(s)F(s) \quad (22)$$

After taking the Laplace transforms of the state space equations, the transfer matrix ( $G(s)$ ) can be derived as:

$$G(s) = C[(sI - A)^{-1}B] + D \quad (23)$$

Hence, the plat transfer functions are then given by:

$$G_1(s) = \frac{x_r(s)}{F(s)} = \frac{A}{D}, G_2(s) = \frac{x_g(s)}{F(s)} = \frac{B}{D}, G_3(s) = \frac{x_w(s)}{F(s)} = \frac{C}{D} \quad (24)$$

where,

$$A = (k_g k_s + k_g k_w + k_s k_w + b_g k_s S + b_s k_g S + b_g k_w S + b_w k_g S + b_s k_w S + b_w k_s S + b_g b_s S^2 + b_g b_w S^2 + b_s b_w S^2 + b_g m_g S^3 + b_g m_r S^3 + b_g m_w S^3 + b_w m_g S^3 + b_s m_w S^3 + k_g m_g S^2 + k_g m_w S^2 + k_w m_g S^2 + k_s m_w S^2 + m_g m_w S^4)$$

$$B = ((k_s + S b_s)(k_g + k_w + b_g S + b_w S + m_w S^2))$$

$$C = ((k_g + S b_g)(k_s + S b_s))$$

$$D = (k_g k_s k_w + b_g b_r b_s S^3 + b_g b_r b_w S^3 + b_g b_s b_w S^3 + b_r b_s b_w S^3 + b_g b_r k_s S^2 + b_r b_s k_g S^2 + b_g b_r k_w S^2 + b_r b_w k_g S^2 + b_g b_s k_w S^2 + b_s b_w k_g S^2 + b_r b_s k_w S^2 + b_r b_w k_s S^2 + b_g b_r m_g S^4 + b_g b_s m_g S^4 + b_g b_s m_r S^4 + b_g b_r m_w S^4 + b_g b_w m_r S^4 + b_r b_w m_g S^4 + b_g b_s m_w S^4 + b_s b_w m_g S^4 + b_r b_s m_w S^4 + b_s b_w m_r S^4 + b_r k_g m_g S^3 + b_g k_s m_g S^3 + b_s k_g m_r S^3 + b_s k_g m_r S^3 + b_g k_w m_r S^3 + b_r k_g m_w S^3 + b_r k_w m_g S^3 + b_w k_g m_r S^3 + b_s k_g m_w S^3 + b_s k_w m_g S^3 + b_w k_s m_g S^3 + b_r k_s m_w S^3 + b_s k_w m_r S^3 + b_w k_s m_r S^3 + b_g m_g m_r S^5 + b_g m_r m_w S^5 + b_r m_g m_w S^5 + b_w m_g m_r S^5 + b_s m_g m_w S^5 + b_s m_r m_w S^5 + k_g k_s m_g S^2 + k_g k_s m_r S^2 + k_g k_w m_r S^2 + k_g k_s m_w S^2 + k_s k_w m_g S^2 + k_s k_w m_r S^2 + k_g m_g m_r S^4 + k_g m_r m_w S^4 + k_w m_g m_r S^4 + k_s m_g m_w S^4 + k_s m_r m_w S^4 + m_g m_r m_w S^6 + b_r k_g k_s S + b_r k_g k_w S + b_g k_s k_w S + b_s k_g k_w S + b_w k_g k_s S + b_r k_s k_w S)$$

As mentioned previously, it is complicated to estimate some unknown model parameters i.e. mass, stiffness and damping parameters in the TMS dynamic equations. For an instant, impedance characteristic of human is normally affected by environment activities, and, and this leads to rough parameter estimation [12]. The MATLAB system identification was consequently adopted, which explains in the following section.

#### Robotic TMS system identification and its implementation

System identification is a method to mathematically construct dynamic characteristic models using test input and output information. The estimated model is always required to be implemented in controlling the plant, designing and synthesizing a controller for the plant, simulating the controlled plant [13-14]. It can be first assumed that the robotic-assisted TMS is modelled as a black box with unknown model parameters. Several parametric model structures describing system transfer functions and its differential equations are available to assist in developing an unknown parameter model. The auto-regressive moving average model with an exogenous signal (ARMAX) is one of the most robust techniques in the parametric modelling of time series used for studying complex dynamical systems. ARMAX was selected since it has more flexibility in the handling of dynamic disturbance modelling, and is suitable for parameter estimation of nonlinear or non-convex problems [15].

The Z-transform is reasonable for analysis of practical model identification. It can be said that the relationship between the eigenvalues and the poles of the ARMAX-based estimated model can be developed beyond the basics that the eigenvalues can be determined through the poles of the model. The ARMAX model structure can be presented in the following equation [19]. The current observation output:  $y(k)$  is the sum of its own past, and the system input is defined as  $u(k)$ . The noise signal moderated into the plant is  $e(k)$ . The structural model equation representing the relationship between the system inputs can be expressed as:

$$y(t) + a_1 y(t-1) + \dots + a_{n_a} y(t-n_a) = b_1 u(t-1) + \dots + b_{n_b} u(t-n_b) + e(t) + c_1 e(t-1) + \dots + c_{n_c} e(t-n_c) \quad (25)$$

$$A(z)y(t) = B(z)u(t) + C(z)e(t) \quad (26)$$

where,  $A(z)$ ,  $B(z)$  and  $C(z)$  are polynomial equations and can be derived from:

$$\begin{aligned}
 A(z) &= 1 + a_1 z^{-1} + \dots + a_{n_a} z^{-n_a} \\
 B(z) &= b_0 + b_1 z^{-1} + \dots + b_{n_b} z^{-n_b} \\
 C(z) &= 1 + c_1 z^{-1} + \dots + a_{n_c} z^{-n_c}
 \end{aligned} \tag{27}$$

The Matlab Identification Toolbox was executed to appropriately estimate the model unknown parameters by selecting a set of a polynomial of  $n_a$ ,  $n_b$  and  $n_c$  in order to minimize the forecasting errors and to obtain the effective model. In addition, to present the system-parameter estimation technique under consideration adequately, the model validation finally verifies a proposed identified model. According to the ARMAX implementation, a set of preliminary experiments involving a robot-assisted TMS system has been developed and carried out.

The main objective of the experiments is to evaluate the dynamic responses of the contact force between the stimulating coil and the human head in real-time. The key parameters are captured in real-time based sampling time of 0.004 s, which are made up of the interactive force setpoint of 2 N (ARMAX input) and the measured contact force (ARMAX output). The test information collected were initially normalized before being appropriately identified their relationship using Matlab Identification Toolbox based on ARMAX command. The discrete-time ARMAX model based on Equation (30) was identified, and its results are given as:

$$\begin{aligned}
 A(z) &= 1 - 0.3855z^{-1} - 0.6006z^{-2} \\
 B(z) &= 0.008147z^{-1} + 0.005766z^{-2} \\
 C(z) &= 1 + 0.1275z^{-1} + 0.03972z^{-2}
 \end{aligned} \tag{28}$$

Model quality validation has been executed in the final stage of the model estimation process in order to verify the simulated model outputs and the actual outcomes. Three model validation techniques were then utilized consisting of the percentage of best fit ( $R^2$ ), the mean sum of square error (MSE) and the final prediction error (FPE). Best fit validation is computed by the determination coefficient  $R^2$  and can be analysed in Equation (33), and MSE is formulated in Equation (34) Additionally, FPE equation can be expressed in Equation (35), to be noted that most estimated models with high accuracy has smallest FPE. After implementing the identification-based ARMAX model, it can be concluded that fit to estimation data is 97.10 %, MSE of 0.00024 was achieved and FPE value is 0.00025. This indicates the actual and estimated data have high similarity in terms of the system responses. Thus, it can be postulated that the acceptable percentage of the model fitting estimation is more than 80% [16].

$$R^2 = 100 \times \left( 1 - \frac{\sum_{i=1}^N \varepsilon^2}{\sum_{i=1}^N (y - \hat{y})^2} \right) \% \tag{29}$$

$$MSE = \sum_{i=1}^N \frac{(y - \hat{y})^2}{N} \tag{30}$$

$$FPE = \frac{1}{N} \sum_{i=1}^N (y - \hat{y})^2 \left[ \frac{(1 + \frac{P}{N})}{(1 - \frac{P}{N})} \right] \tag{31}$$

where,  $y$  and  $\hat{y}$  are actual and calculated outputs,

$N$  is a number of experimental data, and

$P$  is a number of parameters ( $n_a + n_b + n_c$ ).



Regarding the computing results of the ARMAX model, the system transfer functions without noise  $C(z)$  ( $ARMAX_1$ ) and with noise  $C(z)$  ( $ARMAX_2$ ) are as follows:

$$ARMAX_1 = \frac{0.008147z^{-1} + 0.005766z^{-2}}{1 - 0.3855z^{-1} - 0.6006z^{-2}} \quad (32)$$

$$ARMAX_2 = \frac{0.01561 + 0.00199z^{-1} + 0.0006201z^{-2}}{1 - 0.3855z^{-1} - 0.6006z^{-2}} \quad (33)$$

MATLAB discrete-time to continuous-time model conversion was applied to convert the transfer function  $ARMAX_2$  into a continuous-time transfer function ( $ARMAX_{TF}$ ) as in Equation (38), and this is used in the robotic PI force control.

$$ARMAX_{TF} = \frac{0.01561s^3 + 6.541s^2 + 9321s + 1.802e06}{s^3 + 252.7s^2 + 6.331e05s + 1.371e06} \quad (34)$$

### Robotic control implementation for TMS

#### Implementation and results of robotic implicit force control based on PI algorithm

To deliver more reliability and stability in the robotic TMS, implicit position-based force control, which is a principle method in the achievement of the robot's real-time path control, was implemented. Practically, it is suited to control commercial robots as they have been designed as positioning devices. The type of control provides the force control rule to respond simultaneously to the interactive force ( $F_s$ ) between the human head and robot end effector. It can compensate for positioning variations in the robot manipulator. As suggested that the position-based force control is appropriately associated with simple control schemes, in this study, it was hence decided to apply PI control to provide the smallest possible error, since the derivative term in PID is too sensitive to noise, and this could introduce a robot destabilizing effect.

For better understanding, consider the control block diagram as depicted in Figure 1. The  $X$  and  $F$  are  $6 \times 1$  vectors representing the Cartesian position and orientation and Cartesian force and moment respectively. The  $S$  is a  $6 \times 6$  diagonal selection matrix, where each element becomes a one for position control or zero for no position control. The input of PI force control ( $\Delta F$ ) is the difference between the desired force ( $F_{ref}$ ) and  $F_s$ . Each output from the force control ( $X_0^F$ ) yields an equivalent position ( $X_s$ ), which is directly modified by an initial reference position ( $X_0^P$ ). An incremental position ( $\Delta X$ ) can be computed by subtracting  $X_s$  from a current robot position ( $X_0$ ). Remarkably, the optimized PI gains selected in the force control was appropriately defined using PI control design based on Ziegler–Nichols tuning method in MATLAB based on the proposed transfer function in Equation (38). This strives to keep the force overshoot response to a very minimum and to provide rapid response at the same time. However, PI control has a nonlinear effect, namely integral windup, which has to be considered. This occurs when there is a large change in a setpoint, and the Integral term responds to accumulated errors. It results in excessive overshoot from the setpoint value, and to overcome this limitation, the Integral anti-windup was implemented using a discrete-time equation with sampling time  $\tau$ , the time interval  $k$ , PI control output  $U(k)$  and its error  $e(k)$  as:

$$U(k) = U(k - 1) + k_p[e(k) - e(k - 1)] + k_i e(k) \quad (35)$$

MATLAB Simulink block diagram of the robotic TMS system along with PI position-based force control was developed as shown in Figure 4 [17]. Correspondingly, Ziegler–Nichols tuning method, which has been widely used because of their simplicity, deliveries optimized proportional ( $k_p$ ) and integral ( $k_i$ ) gains of 1.176 and 2.850 respectively. The test results, where a system setpoint was fixed at the contact force of 2 N, display in Figure 5. This yields the rise time, settling time and overshoot performance of 0.596 s, 0.957 s and 0.386 %. As the frequency response was examined, then a closed-loop stability analysis was applied, and the evaluation revealed satisfactorily stable performance with a phase margin of 88.00 degrees at 3.47 rad/s. It can be summarized that

the set of conveyed PI gains demonstrates significant satisfaction with the overall control performance in the robotic TMS system, and this is subsequently used in the robot implicit force control approach.

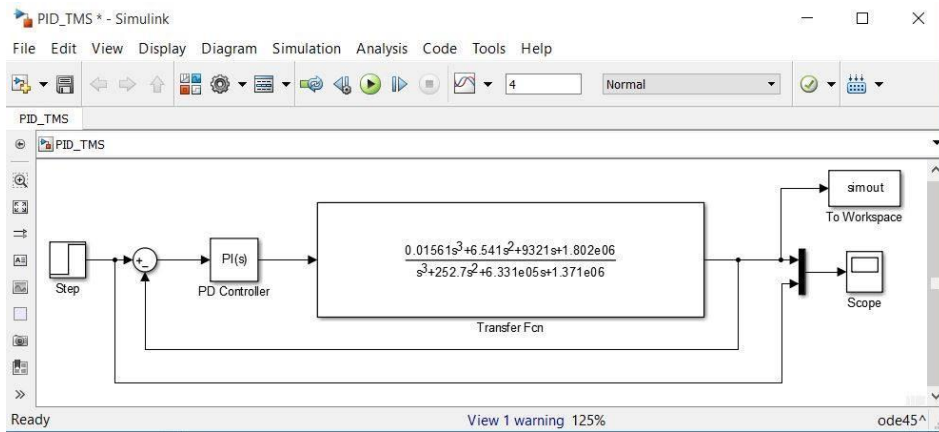


Figure 4. MATLAB simulation diagram

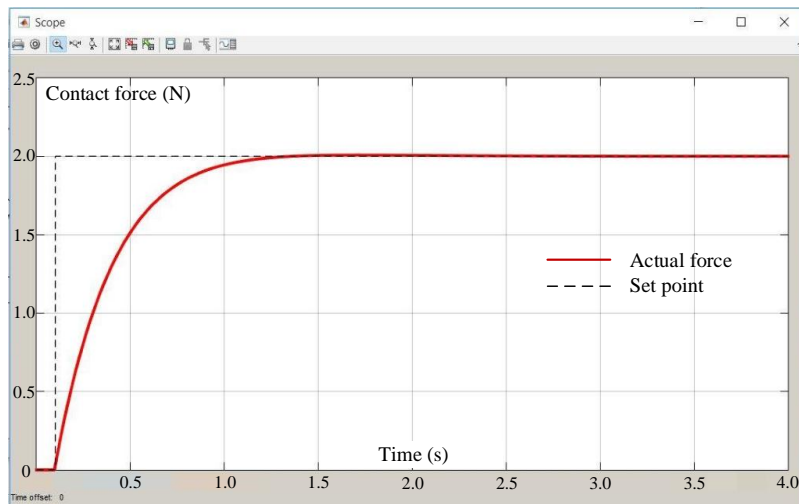


Figure 5. MATLAB simulation results

#### Safety Issues in Robot-assisted TMS

The most crucial issue in the robotic TMS system is safety because any failures occurring might become very serious problems. Hardware and software safety issues have been given due consideration. Stand-alone emergency stop and manual control panel (MCP) are available in order to manually deactivate the system by both administrator and patient if accidents are detected. The speed of the end effector was limited to 10 mm/s. A timeout was assigned for the serial and TCP/IP communication. If the transmission has not been completed yet after the timeout has elapsed, then all communication will be immediately terminated. Safety threshold forces ( $\hat{F}_x$ ,  $\hat{F}_y$  and  $\hat{F}_z$ ) were individually defined. If one or more interactive forces are greater than the proposed threshold values, then the robot will be suddenly stopped and waiting for the reset command in order to ensure the safety of the human participant.

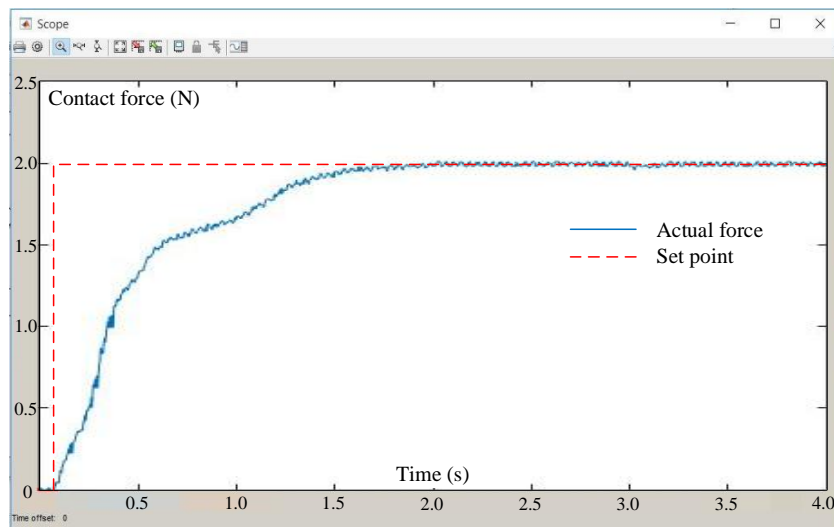
#### Results of Proposed Robotic Force Control based on the PI Control Approach

The implementation of the optimized PI gains for the robot force control was successfully executed. The qualitative evaluation of the control approach is then carried out. The experimental procedure begins with initially setting the fixed contact force of 2 N as suggested. The robot end effector attached by the TMS coil is moving to the human's head position with appropriate orientation using the 3D tracking camera. After completely approaching the target, the robotic implicit force control based on the proposed PI control algorithm is simultaneously activated. In the meantime, interactive force is continuously measured respective to time. Figure

6 provides the results obtained from the real test. Root Mean Square Error (RMSE) presents statistically the comparison between two data sets. RMSE can be formulated as the following equation, where  $y$  and  $\hat{y}$  are the actual and desired contact force data, and  $N$  is a number of captured experimental data.

$$RMSE = \sqrt{\sum_{i=1}^N \frac{(y - \hat{y})^2}{N}} \quad (36)$$

The RMSE technique was implemented to evaluate the qualitative performance, and it gives 0.014 N. It means that the real interactive force exhibits to track closely the desired input with approximately zero error at the steady state. Notably, the overshoot response is not presented, that means the controlled robot system has the capability to modify the interactive force without any higher force amplitude applied to the patient’s head. It can also protect the human subject from the risk of harm or injury by the KUKA robot. Therefore, this PI control algorithm is considered acceptable for the effective robot-assisted TMS system, and it offers the safe and reliable haptic interaction between the patient and the KUKA robot manipulator.



**Figure 6.** Results obtained from the Robotic TMS system

## CONCLUSION

This paper initially extends the dynamic model of the robot-assisted TMS system and its transfer function; however, estimating the relevant impedance parameters has encountered a problematic aspect by reason of dealing with the complicated dynamic characteristic of the human behaviour and the TMS system. In order to overcome this challenging issue, MATLAB model identification using ARMAX model was utilized to facilitate the appropriately approximated discrete-time and continuous-time transfer functions of the robotic TMS system. The model validation was summarized that the percentage of best fit of 97.10 % was achieved, and the FPE and MSE outcomes based on the smallest values are 0.0002, and of 0.0002 was achieved indicating that calculated mathematical models are a highly effective and reliable model.

As the ARMAX model has the capability to effectively identify the TMS plant transfer function, subsequently the PI force control algorithm was simulated based on Ziegler–Nichols gains-tuning method. It strategically offers the optimized  $k_p$  and  $k_i$  gains of 1.176 and 2.850 respectively. The gain parameters produced was used in the robotic implicit PI force control. The test results of the real robotic TMS reported that this provides an effective matching performance between the setpoint and actual outcomes, which is proved by the RMSE value of 0.014 N at the steady state. It can be concluded that as the quantitative measurement of the robotic control approach was validated and demonstrates effective performance and stability. Therefore the proposed robotic Transcranial magnetic stimulation using PI implicit force control is significantly considered acceptable for the TMS system.

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